INTEGRATED CIRCUITS

DATA SHEET

TDA1023/T

Proportional-control triac triggering circuit

Product specification Supersedes data of August 1982 IC02 May 1991

PEATURES

APPLICATIONS

GENERAL DESCRIPTION

Philips Semiconductors



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TDA1023/T

FEATURES

- Adjustable width of proportional range
- Adjustable hysteresis
- · Adjustable width of trigger pulse
- Adjustable repetition timing of firing burst
- · Control range translation facility
- Fail safe operation
- · Supplied from the mains
- Provides supply for external temperature bridge

APPLICATIONS

- Panel heaters
- Temperature control

GENERAL DESCRIPTION

The TDA1023 is a bipolar integrated circuit for controlling triacs in a proportional time or burst firing mode. Permitting precise temperature control of heating equipment it is especially suited to the control of panel heaters. It generates positive-going trigger pulses but complies with regulations regarding mains waveform distortion and RF interference.

QUICK REFERENCE DATA

SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
Vcc	supply voltage (derived from mains voltage)	-	13.7	-	V
Vz	stabilized supply voltage for temperature bridge	-	8	-	V
116(AV)	supply current (average value)	-	10	-	mA
t _w	trigger pulse width	-	200	-	μs
Ть	firing burst repetition time at C_T = 68 μ F	-	41	-	s
-I _{OH} *	output current	-	-	150	mA
T _{amb}	operating ambient temperature range	-20	-	+75	°C

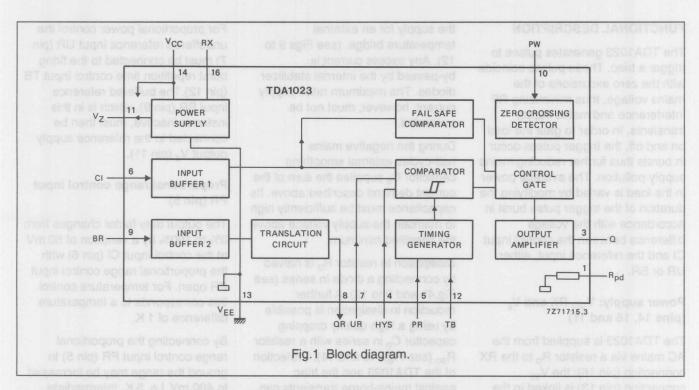
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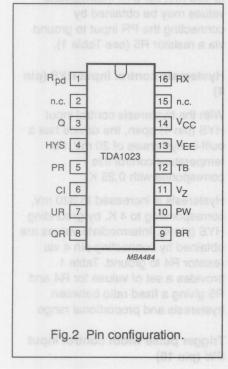
*Negative current is defined as conventional current flow out of a device. A negative output current is suited for positive triac triggering.

ORDERING INFORMATION

EXTENDED	PACKAGE					
TYPE NUMBER	PINS	PIN POSITION	MATERIAL	CODE		
TDA1023	16	DIL	plastic	SOT38		
TDA1023T	16	mini-pack	plastic	SO16; SOT109A		

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PINNING

SYMBOL	PIN	DESCRIPTION
R _{pd}	1	internal pull-down resistor
n.c.	2	not connected
Q	3	output
HYS	4	hysteresis control input
PR grade vel ber	50 el eonamio	proportional range control input
CI	6	control input
UR	7	unbuffered reference input
QR	8	output of reference buffer
BR botellu	or 9 wit of batosa	buffered reference input
PW BIRT TYRE	10	pulse width control input
Vz	11	reference supply output
TB and h	12	firing burst repetition time control input
V _{EE}	13	ground lamence entirippoint themus er
Vcc	14	positive supply
n.c.	15	not connected
RX	16	external resistor connection

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FUNCTIONAL DESCRIPTION

The TDA1023 generates pulses to trigger a triac. These pulses coincide with the zero excursions of the mains voltage, thus minimizing RF interference and mains supply transients. In order to gate the load on and off, the trigger pulses occur in bursts thus further reducing mains supply pollution. The average power in the load is varied by modifying the duration of the trigger pulse burst in accordance with the voltage difference between the control input CI and the reference input, either UR or BR.

Power supply: V_{cc}, RX and V_z (pins 14, 16 and 11)

The TDA1023 is supplied from the AC mains via a resistor R_D to the RX connection (pin 16); the V_{EE} connection (pin 13) is linked to the neutral line (see Fig.4a). A smoothing capacitor C_S should be coupled between the V_{CC} and V_{EE} connections.

A rectifier diode is included between the RX and $V_{\rm CC}$ connections whilst the DC supply voltage is limited by a chain of stabilizer diodes between the RX and $V_{\rm EE}$ connections (see Fig.3).

A stabilized reference voltage (V_z) is available at pin 11 to power an external temperature sensing bridge.

Supply operation

During the positive mains half-cycles the current through the external voltage dropping resistor R_D charges the external smoothing capacitor C_S until RX attains the stabilizing potential of the internal stabilizing diodes. R_D should be selected to be capable of supplying the current I_{CC} for the TDA1023, the average output current $I_{3(AV)}$, recharge the smoothing capacitor C_S and provide

the supply for an external temperature bridge. (see Figs 9 to 12). Any excess current is by-passed by the internal stabilizer diodes. The maximum rated supply current, however, must not be exceeded.

During the negative mains half-cycles external smoothing capacitor C_s supplies the sum of the current demand described above. Its capacitance must be sufficiently high to maintain the supply voltage above the specified minimum.

Dissipation in resistor $R_{\rm D}$ is halved by connecting a diode in series (see Fig.4b and 9 to 12). A further reduction in dissipation is possible by using a high quality dropping capacitor $C_{\rm D}$ in series with a resistor $R_{\rm SD}$ (see Figs 4c and 14). Protection of the TDA1023 and the triac against mains-borne transients can be provided by connecting a suitable VDR across the mains input.

Control and reference inputs CI, BR and UR (pins 6, 9 and 7)

For the control of room temperature (5 °C to 30 °C) optimum performance is obtained by using the translation circuit. The buffered reference input BR (pin 9) is used as a reference input whilst the output reference buffer QR (pin 8) is connected to the unbuffered reference input UR (pin 7). This ensures that the range of room temperature is encompassed in most of the rotation of the potentiometer to give a linear temperature scale with accurate setting.

Should the translation circuit not be required, the unbuffered reference input UR (pin 7) is used as a reference input. The buffered reference input BR (pin 9) must then be connected to the reference supply output V_Z (pin 11).

For proportional power control the unbuffered reference input UR (pin 7) must be connected to the firing burst repetition time control input TB (pin 12). The buffered reference input BR (pin 9), which is in this instance inactive, must then be connected to the reference supply output V_{z} (pin 11).

Proportional range control input PR (pin 5)

The output duty factor changes from 0% to 100% by a variation of 80 mV at the control input CI (pin 6) with the proportional range control input PR open. For temperature control this corresponds to a temperature difference of 1 K.

By connecting the proportional range control input PR (pin 5) to ground the range may be increased to 400 mV, i.e. 5 K. Intermediate values may be obtained by connecting the PR input to ground via a resistor R5 (see Table 1).

Hysteresis control input HYS (pin 4)

With the hysteresis control input HYS (pin 4) open, the device has a built-in hysteresis of 20 mV. For temperature control this corresponds with 0.25 K.

Hysteresis is increased to 320 mV, corresponding to 4 K, by grounding HYS (pin 4). Intermediate values are obtained by connecting pin 4 via resistor R4 to ground. Table 1 provides a set of values for R4 and R5 giving a fixed ratio between hysteresis and proportional range.

Trigger pulse width control input PW (pin 10)

The width of the trigger pulse may be adjusted to the value required for the triac by choosing the value of the external synchronization resistor

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R_s between the trigger pulse width control input PW (pin 10) and the AC mains. The pulse width is inversely proportional to the input current (see Fig.13).

Output Q (pin 3)

Since the circuit has an open-emitter output it is capable of sourcing current. It is thus suited for

generating positive-going trigger pulses. The output is current-limited and short-circuit protected. The maximum output current is 150 mA and the output pulses are stabilized at 10 V for output currents up to that value.

To minimize the total supply current and power dissipation, a gate resistor R_G must be connected

between the output Q and the triac gate to limit the output current to the minimum required by the triac (see Figs 5 to 8).

Pull-down resistor Rpd (pin 1)

The TDA1023 includes a $1.75~\text{k}\Omega$ pull-down resistor R_{pd} between pins 1 and 13 (V_{EE} , ground connection) intended for use with sensitive triacs.

LIMITING VALUES

In accordance with the Absolute Maximum System (IEC 134)

SYMBOL	PARAMETER	MIN.	MAX.	UNIT
Vcc	DC supply voltage	-	16	V
Supply curren	t	e a selet till bas	CIES Proportion	nanday han laster
I _{16(AV)}	average	· ·	30	mA
I _{16(RM)}	repetitive peak	2 Indino	100	mA
1 _{16(SM)}	non-repetitive peak (tp < 50 μs)	-	2	A
Vı	input voltage, all inputs	-	16	V
l _{6, 7, 9, 10}	input current		10	mA
V ₁	voltage on R _{pd} connection	9 -	16	V
V _{3. 8. 11}	output voltage, Q, QR, Vz	- (e a/d)	16	V
Output curren	1 08 08 nago a r	ng	egnat lange	AV _e proportion
-I _{OH(AV)}	average	191	30	mA
-I _{OH(M)}	peak max. 300 µs	. (700	mA
P _{tot}	total power dissipation	ell -	500	mW
T _{stg}	storage temperature range	-55	+150	°C Maries Rounies
T _{amb}	operating ambient temperature range	-20 el oller	+75	°C

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CHARACTERISTICS

 V_{CC} = 11 to 16 V; T_{amb} = -20 to +75 °C unless otherwise specified

SYMBOL	PARAMETER Am 08 rs m	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply	Desilidate or to the transfer of the contract	and the output pulses to				9-19).
Vcc 24 av. t	internally stabilized supply voltage at I ₁₆ = 10mA	value. To minimize the total as	12	13.7	15	V
$\Delta V_{CC}/\Delta I_{16}$	variation with I ₁₆	and power dissipation	- Bi	30	bapabie capabie	mV/mA
I ₁₆	supply current at $V_{16-13} = 11$ to 16 V $I_{10} = 1$ 1mA; $f = 50$ Hz; pin 11 open; $V_{6-13} > V_{7-13}$	pins 4 and 5 open	-	-	6	mA
		pins 4 and 5 grounded	- Mah		7.1	mA
Reference	supply output V _Z (pin 11) for external ten	nperature bridge				
V ₁₁₋₁₃	output voltage output current	11313000	apsilov	8 que 0	1	V mA
	d reference inputs CI, BR and UR (pins 6	, 9 and 7)			709111	o kideni
V ₆₋₁₃	input voltage to inhibit the output input current	V ₁ = 4 V	- ceak	7.6	2	V μA
	control input HYS (pin 4)	(eq 00 > qr)	live peak	dadalulo		(888)
ΔV_6 ΔV_6	hysteresis hysteresis	pin 4 open pin 4 grounded	9	20 320	40	mV mV
	al control range input PR (pin 5)	V an	D one	tour humbs		1
ΔV_6	proportional range	pin 5 open	50	80	130	mV
ΔV_6	proportional range	pin 5 grounded	-	400	1	mV
Pulse widt	h control input PW (pin 10)					
t _w	pulse width	$I_{10(RMS)} = 1mA; f = 50 Hz$	100	200	300	μs
Firing burs	t repetition time control input TB (pin 12	e range (
ТьСт	firing burst repetition time, ratio to capacitor C _T	temperature range	320	600	960	ms/μF
Output of r	reference buffer QR (pin 8)					
V ₈₋₁₃ V ₈₋₁₃ V ₈₋₁₃	output voltage at input voltage:	V ₉₋₁₃ = 1.6 V V ₉₋₁₃ = 4.8 V V ₉₋₁₃ = 8 V		3.2 4.8 6.4		V V
Output Q (pin 3)					
V _{OH}	output voltage HIGH output current HIGH	-I _{OH} = 150 mA	10	-	150	V mA
Internal pu	III-down resistor R _{pd} (pin 1)					
R _{pd}	resistance to V _{EE}		1	1.75	3	kΩ

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Table 1 Adjustment of proportional range and hysteresis. Combinations of resistor values giving hysteresis > 1/4 proportional range.

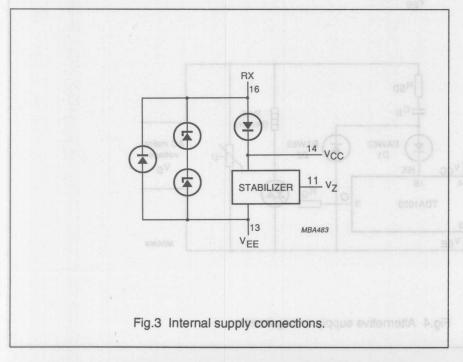
Proportional range	Proportional range resistor	Minimum hysteresis	Maximum hysteresis resistor
	R5		R4
mV	kΩ	mV	kΩ
80	open (netaent)	20	open
160	3.3	40	9.1
240	1.1	60	4.3
320	0.43	80	2.7
400	0	100	1.8

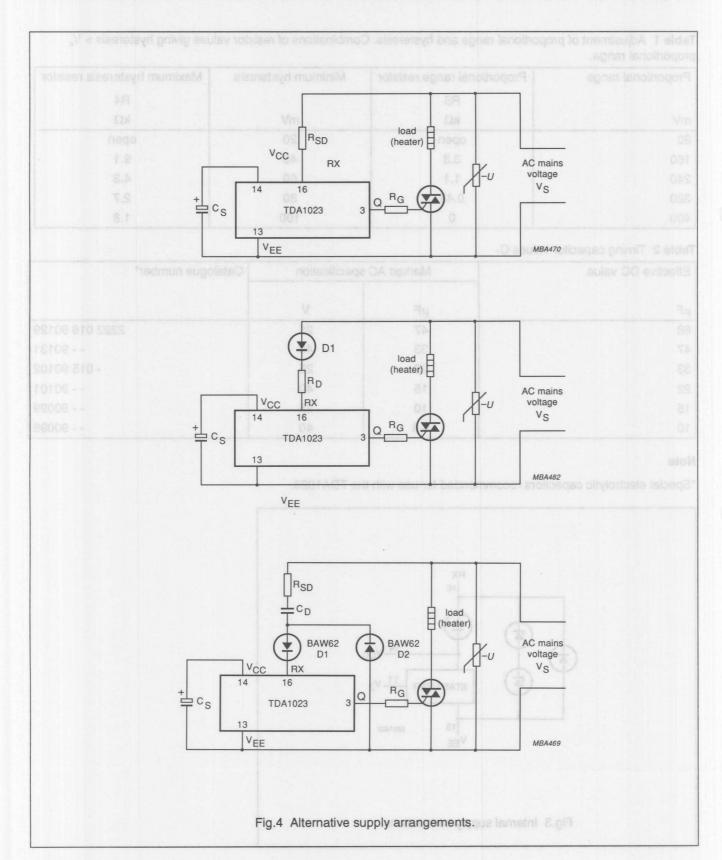
Table 2 Timing capacitor values C_T

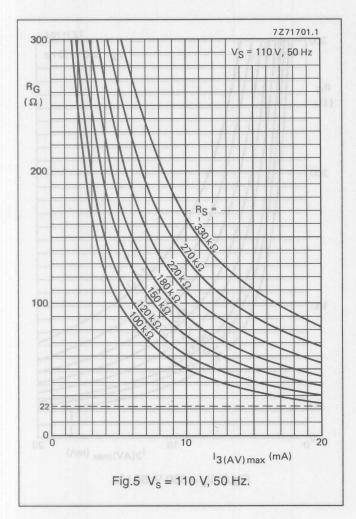
Effective DC value	Marked AC s	pecification	Catalogue number*	
μF	μF	V		
68	47	25	2222 016 90129	
47	33	40	90131	
33	22	25	- 015 90102	
22 entan 3	15	40	90101	
15	10	25	90099	
10	6.8	40	90098	

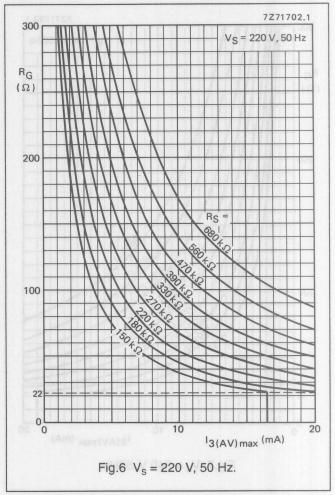
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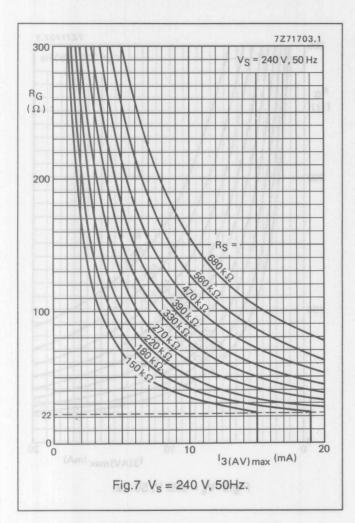
^{*}Special electrolytic capacitors recommended for use with the TDA1023.

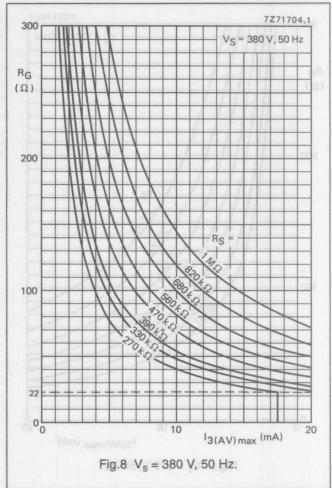


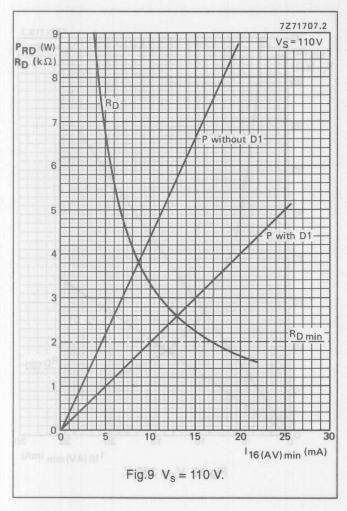


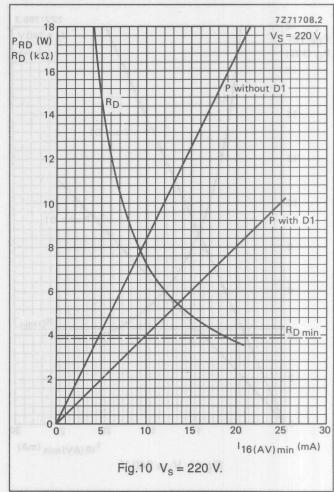


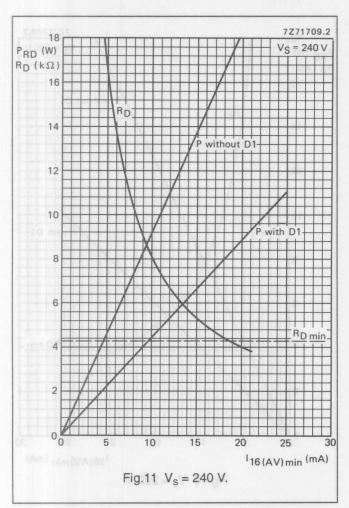


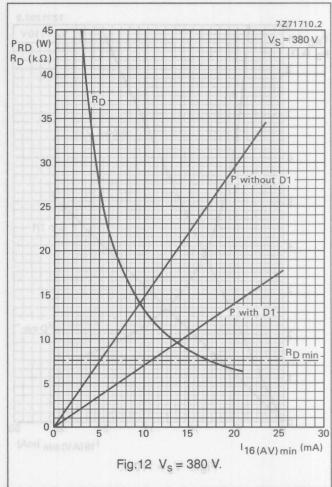


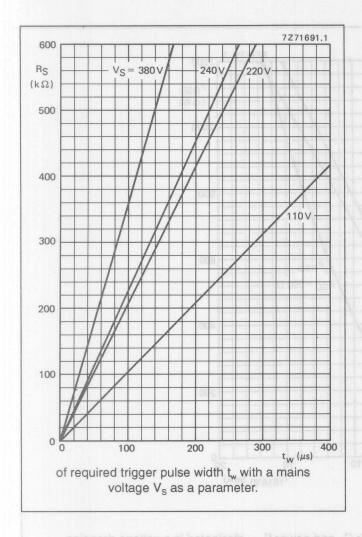












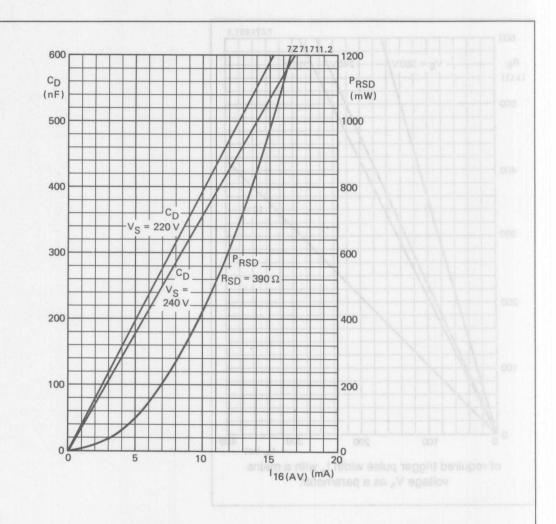
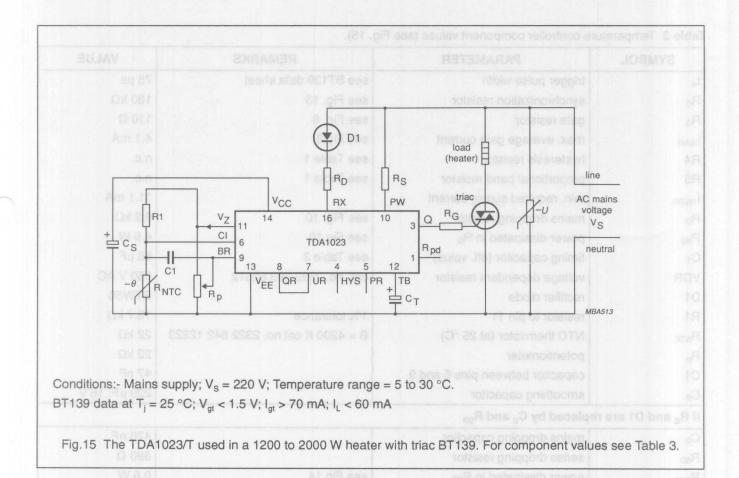


Fig.14 Nominal value of voltage dropping capacitor C_D and power P_{RSD} dissipated in a voltage dropping resistor R_{SD} as a function of average supply current $I_{16~(AV)}$ with the mains supply voltage V_S as a parameter.



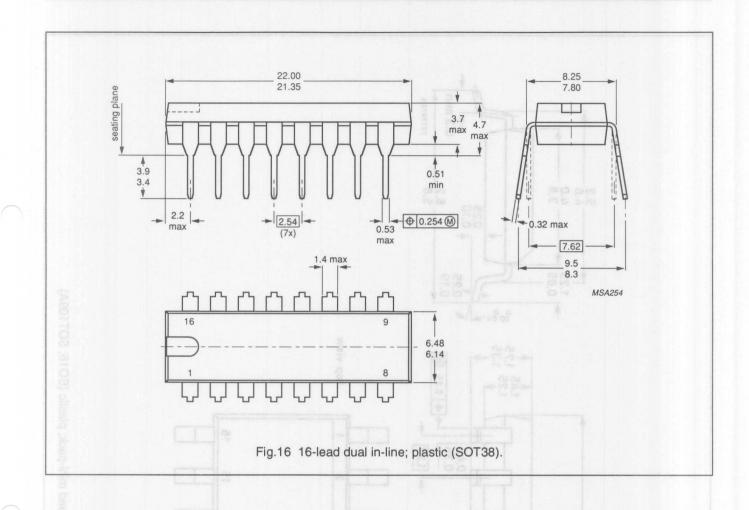
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Table 3 Temperature controller component values (see Fig. 15).

SYMBOL	PARAMETER	REMARKS	VALUE
t _w	trigger pulse width	see BT139 data sheet	75 μs
Rs	synchronization resistor	see Fig. 13	180 kΩ
R _G	gate resistor	see Fig. 6	110 Ω
I _{3(AV)}	max. average gate current	see Fig.8	4.1 mA
R4	hysteresis resistor	see Table 1	n.c.
R5	proportional band resistor	see Table 1	n.c.
I _{16(AV)}	min. required supply current	xa zov	11.1 mA
R _D	mains dropping resistor	see Fig. 10	6.2 kΩ
P _{RD}	power dissipated in R _D	see Fig.10	4.6 W
Ст	timing capacitor (eff. value)	see Table 2	68 μF
VDR	voltage dependent resistor	cat. no. 2322 593 62512	250 V AC
D1	rectifier diode	NIC L'AP	BYW56
R1	resistor to pin 11	1% tolerance	18.7 kΩ
R _{NTC}	NTC thermistor (at 25 °C)	B = 4200 K cat no. 2322 642 12223	22 kΩ
Rp	potentiometer		22 kΩ
C1	capacitor between pins 6 and 9		47 nF
Cs	smoothing capacitor	supply: V _s = 220 V; Temperature range	220 μF; 16 V
If R _D and D1 are	replaced by C _D and R _{SD}	100 - J. William P. Le Co. C.	The may see it
C _D	mains dropping capacitor	COST WEST IN A TONO IN SOUR WITHOUT	470 nF
R _{SD}	series dropping resistor	3 3 5 7 13 5 5 5 5 5 5 6 6 6 6 6 6 6 6 6 6 6 6 6	390 Ω
P _{RSD}	power dissipated in R _{SD}	see Fig.14	0.6 W
VDR	voltage dependent resistor	cat. no. 2322 594 62512	250 V AC

Notes

- 1. ON/OFF control: pin 12 connected to pin 13.
- 2. If translation circuit is not required: slider of R_p to pin 7; pin 8 open; pin 9 connected to pin 11.



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SOLDERING

Plastic dual-in line packages

BY DIP OR WAVE

The maximum permissible temperature of the solder is 260 °C. This temperature must not be in contact with the joint for more than 5 s. The total contact time of successive solder waves must not exceed 5 s.

The device may be mounted up to the seating plane but the temperature of the plastic body must not exceed the specified storage maximum. If the printed-circuit board has been preheated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

REPAIRING SOLDERED JOINTS (BY HAND)

Apply the soldering iron below the seating plane (or not more than 2 mm above it). If its temperature is below 300 °C it must not be in contact for more than 10 s; if between 300 and 400 °C, for not more than 5 s.

Plastic mini-packs

BY WAVE

During placement and before soldering, the component must be fixed with a droplet of adhesive. After curing the adhesive, the component can be soldered. The

adhesive can be applied by screen printing, pin transfer or syringe dispensing.

Maximum permissible solder temperature is 260 °C and maximum duration of package immersion in solder bath is 10 s, if allowed to cool to less than 150 °C within 6 s. Typical dwell time is 4 s at 250 °C.

A modified wave soldering technique is recommended using two solder waves (dual wave) in which a turbulent wave with high upward pressure is followed by a smooth laminar wave. Using a mildly-activated flux eliminates the need for removal of corrosive residues in most applications.

BY SOLDER PASTE REFLOW

Reflow soldering requires the solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the substrate by screen printing, stencilling or pressure-syringe dispensing before device placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt, infrared, and vapour-phase reflow. Dwell times vary between 50 and 300 s according to method. Typical reflow temperatures range from 215 to 250 °C.

Preheating is necessary to dry the paste and evaporate the binding

agent. Preheating duration: 45 min at 45 °C.

REPAIRING SOLDERED JOINTS (BY HAND-HELD SOLDERING IRON OR PULSE-HEATED SOLDER TOOL)

Fix the component by first soldering two, diagonally opposite, end pins. Apply the heating tool to the flat part of the pin only. Contact time must be limited to 10 s at up to 300 ° C. When using proper tools, all other pins can be soldered in one operation within 2 to 5 s at between 270 and 320 °C. (Pulse-heated soldering is not recommended for SO packages).

For pulse-heated solder tool (resistance) soldering of VSO packages, solder is applied to the substrate by dipping or by an extra thick tin/lead plating before package placement.

CLEANING

Avoid cleaning if possible.

If cleaning is necessary use cold or hot water. A detergent may be added to the water. Finally rinse with de-ionized water.

Do **not** use ultrasonic cleaning methods as these may damage the inner or outer leads.

Do not use solvents.

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DEFINITIONS

This data sheet containstarget or goal specifications for product development.
This data sheet contains preliminary data; supplementary data may be published later.
This data sheet contains final product specifications.

Limiting values

Limiting values are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those given in the Characteristics sections of this specification is not implied. Exposure to limiting values for extended periods may affect device reliability.

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